# **Optimization of Nano-Scaled Implantation Free InGaAs MOSFETs**

Roland Stenzel<sup>(1)</sup>, Wilfried Klix<sup>(1)</sup>, Tom Herrmann<sup>(1)</sup>, Stefan Dünkel<sup>(1)</sup>, Ralf Illgen<sup>(1)</sup>

<sup>(1)</sup>University of Applied Sciences Dresden Friedrich-List-Platz 1 01069 Dresden Germany Email: stenzel@et.htw-dresden.de

### Abstract

The influence of structure parameters of implantation free InGaAs MOSFETs on the static and dynamic transistor behavior was simulated by a quantum hydrodynamic model calibrated by experimental results from a 1  $\mu$ m transistor. Structures with gate lengths from 100 nm to 25 nm were investigated. For the state-of-the-art gate length of 50 nm, cut-off frequencies  $f_t = 230$  GHz and  $f_{max} = 350$  GHz were calculated.

Keywords: InGaAs MOSFET, enhancement mode, simulation, quantum hydrodynamic model.

### **1. INTRODUCTION**

Due to the higher electron mobility in GaAs and other III-V compound semiconductor materials, GaAs-based devices like MESFETs and HEMTs are widely used for high speed circuit applications and monolithic microwave integrated circuits. However, the gate current of the Schottky contact increases strongly at forward biased gate contacts. Therefore enhancement-mode MESFETs or HEMTs are strongly limited in their control voltage swing. In contrast, GaAs-based MOSFETs enable larger gate voltages and have also higher input impedances. Hence low-power circuits with high-speed operation can be combined with high packing density.

In [1], [2], [3] an enhancement-mode n-channel MOSFET was presented with an amorphous Ga<sub>2</sub>O<sub>3</sub>/GdGaO dielectric ( $\kappa \approx 20$ ), a gate length of 1 µm and an In<sub>0.3</sub>Ga<sub>0.7</sub>As channel. Based on the experimental results, the model parameters of a quantum hydrodynamic transport model [4], which is implemented in our 2D/3D-numerical device simulation program SIMBA [5], are calibrated. Thereafter the structure was scaled for gate lengths of 100 nm, 50 nm and 25 nm, respectively. Static and dynamic device simulation results are I-V characteristics, transconductances as well as cut-off frequencies.

## 2. SIMULATION MODELS

Quantum hydrodynamic (QHD) models, which are based on a quantum fluid dynamic model, offer new ways to understand and design quantum sized semiconductor devices. The advantage of this model is its macroscopic character, which enables to obtain a description without knowing of quantum mechanical details like the initial wave function [6]. The classical hydrodynamic (HD) model for semiconductor device simulation can be extended by further terms in the transport equations and in the energy balance equations. They describe an internal quantum potential in the transport equation as well as a quantum heat flux in the energy balance equation. These additional terms in the classical hydrodynamic model allow to describe the continuous electron and hole distribution in a semiconductor device, accumulations of carriers in potential wells and resonant tunneling of carriers, respectively. The standard model for universal device simulations is the drift-diffusion (DD) model, which can be derived from the above mentioned model.

Basic equations of the QHD model are the Poisson equation, continuity equations, transport equations, energy balance equations, energy flux density equations and equations for the quantum correction potential. Further approaches are necessary for carrier mobilities, generation and recombination rates, diffusion coefficients and energy relaxation times, which are almost material dependent. Solutions of the equations are achieved by a successive algorithm (the so called Gummel algorithm). For solving the partial differential equations a box method is used. The resulting non-linear equation systems are solved by the NEWTON-method and the corresponding linear equation systems by preconditioned gradient methods. All models are implemented fully three-dimensionally in the SIMBA program system [4], [5].

### **3. BASIC STRUCTURE SIMULATION AND VERIFICATION**

The starting point for the simulations is the basic structure represented in Fig. 1, as a functionally relevant detail of the real device, published in [3].



Fig. 1. Basic structure for the simulations [3]

The different structure parameters are gate length as well as gate-to-source and gate-to-drain distance  $L_G = L_{GS} = L_{GD} =$ 1  $\mu$ m, gate oxide thickness t<sub>ox</sub> = 17 nm, top delta doping N<sub>st</sub> = 0.6 · 10<sup>12</sup> cm<sup>-2</sup> and bottom delta doping N<sub>sb</sub> = 1.8 · 10<sup>12</sup> cm<sup>-2</sup>. All other layers are undoped. The dielectric gate stack consists of Ga2O3/(Gd0.6Ga0.4)2O3 with a dielectric constant  $\kappa = 20$  [3]. Fig. 2 shows the calculated transfer characteristic and the transconductance in comparison to experimental results from [3]. The simulation of the small-signal behavior results in a transit frequency  $f_t = 12$  GHz and a maximum frequency of oscillation  $f_{max} = 32 \text{ GHz}$  (Fig. 2).



simulation and experiment [3]

#### Fig. 3. Calculated small-signal gains

#### **4. PARAMETER VARIATIONS**

In the following, the structure (Fig. 1) was scaled for gate lengths of 100 nm, 50 nm and 25 nm, respectively. The layer structure for different gate lengths [L<sub>G</sub> = 100 nm / 50 nm / 25 nm] consists of a 40 nm undoped Al<sub>0.24</sub>Ga<sub>0.76</sub>As layer, a 5 nm Al<sub>0.2</sub>Ga<sub>0.8</sub>As barrier layer, a bottom Si δ-doping, a [2 nm / 1.5 nm / 1 nm] Al<sub>0.2</sub>Ga<sub>0.8</sub>As spacer layer, a [2 nm / 1.5 nm / 1 nm] GaAs spacer layer, a [ $t_{ch}$  = 10 nm / 5nm / 4 nm] undoped In<sub>0.3</sub>Ga<sub>0.7</sub>As channel layer, a [2 nm / 1.5 nm / 1 nm] GaAs spacer layer, a top Si  $\delta$ -doping, a [2 nm / 1.5 nm / 1 nm] Al<sub>0.45</sub>Ga<sub>0.55</sub>As barrier layer and a t<sub>ox</sub> = 4 nm  $Ga_2O_3/GdGaO$  dielectric stack. Furthermore  $L_{GS} = L_{GD} = L_G$  was assumed for all structures. The threshold voltage was

adjusted with two equal  $\delta$ -dopings  $[2 \cdot 10^{12} \text{ cm}^{-2} / 3 \cdot 10^{12} \text{ cm}^{-2} / 3.2 \cdot 10^{12} \text{ cm}^{-2}]$  to  $V_{th} = 0.1 \text{ V}$  at  $V_{DS} = 2\text{V}$ . Only at  $L_G = 25 \text{ nm}$ , a reduction of gate oxide was not avertable for further improvements, hence  $t_{ox} = 2 \text{ nm}$  was fixed as a lower limit. The calculated transfer characteristics and the transconductance for separate gate lengths are represented in Fig. 4. Fig. 5 shows the cut-off frequencies  $f_t$  and  $f_{max}$  at  $V_{DS} = 2 \text{ V}$  and  $V_{GS}$  at the maximum transconductance value. At  $L_G = 25 \text{ nm}$ , a maximum transconductance of 1510 mS/mm and cut-off frequencies  $f_t = 300 \text{ GHz}$  and  $f_{max} = 380 \text{ GHz}$  can be achieved.



Fig. 4. Transfer characteristics and transconductance for different gate lengths

Fig. 5. Cut-off frequencies for different gate lengths

For the state-of-the-art gate length of 50 nm further simulations have been carried out. The channel thickness  $t_{ch}$  was varied in the range of 4 nm to 8 nm. The threshold voltage was adjusted to  $V_{th} = 0.1$  V by adapted  $\delta$ -dopings. The resulting maximum drain current and transconductance at  $V_{GS} = V_{DS} = 2V$  are depicted in Fig. 6 and the corresponding cut-off frequencies in Fig. 7. Smaller channels lead in connection with higher  $\delta$ -dopings to increasing currents and transconductance and consequently to increasing cut-off frequencies.



400  $V_{DS} = 2V, V_{GS}@g_{m,max}$  $V_{th} = 0.1V, L_{G} = 50nm$ 350 300 f<sub>t</sub>, f<sub>max</sub> (GHz) 250 f<sub>t</sub> 200 150 100 4 5 6 7 8 3 q t<sub>ch</sub> (nm)

Fig. 6. Maximum drain current and transconductance for different channel thickness

Fig. 7. Cut-off frequencies for different channel thickness

The results of oxide thickness variation are represented in Fig. 8 and Fig. 9. Thin oxides allow higher dopings at the same threshold voltage and thus higher drain currents and transconductances. Reduced oxide thickness causes a rising gate capacity and therefore the increasing transconductance results only in a moderate increment of cut-off frequencies. A structure with  $L_G = 50$  nm and  $t_{ch} = 4$  nm as well as  $t_{ox} = 2$  nm achieves  $g_{m,max} = 1730$  mS/mm,  $f_t = 230$  GHz and  $f_{max} = 350$  GHz.



Fig. 8. Maximum drain current and transconductance for different gate oxide thickness



Fig. 9. Cut-off frequencies for different gate oxide thickness

#### **5. CONCLUSIONS**

Numerical simulations of nano-scaled implantation free InGaAs MOSFETs have been carried out by a quantum hydrodynamic model. After model parameter calibration by experimental results the structure was scaled down to 25 nm gate length. Structure parameter variations are carried out for the state-of-the-art gate length of 50 nm. In this case cut-off frequencies of  $f_t = 230$  GHz and  $f_{max} = 350$  GHz can be reached.

### 6. REFERENCES

- [1] M. Passlack, K. Rajagopalan, J. Abrokwah, and R. Droopad, "Implantation-Free High-Mobility Flatband MOSFET: Principles of Operation," *IEEE Trans. on Electron Devices*, vol. 53, pp. 2454-2459, October 2006.
- [2] K. Rajagopalan, J. Abrokwah, R. Droopad, and M. Passlack, "Enhancement-Mode GaAs n-Channel MOSFET," *IEEE Electron Device Letters*, vol. 27, pp. 959-962, December 2006.
- [3] K. Rajagopalan, R. Droopad, J. Abrokwah, P. Zurcher, P. Fejes, and M. Passlack, "1-µm Enhancement Mode GaAs N-Channel MOSFETs With Transconductance Exceeding 250 mS/mm," *IEEE Electron Device Letters*, vol. 28, pp. 100-102, February 2007.
- [4] J. Höntschel, R. Stenzel, and W. Klix, "Simulation of Quantum Transport in Monolithic ICs Based on In<sub>0.53</sub>Ga<sub>0.47</sub>As-In<sub>0.52</sub>Al<sub>0.48</sub>As RTDs and HEMTs With a Quantum Hydrodynamic Transport Model," *IEEE Trans. on Electron Devices*, vol.51, pp. 684-692, May 2004.
- [5] W. Klix, and R. Stenzel, "SIMBA-User Manual," http://www.htw-dresden.de/~klix/simba/welcome.html.
- [6] A. Wettstein, A. Schenk, and W. Fichtner, "Quantum Device Simulation With the Density-Gradient Model on Unstructured Grids," *IEEE Trans. on Electron Devices*, vol. 48, pp. 279-284, February 2001.

#### ACKNOWLEDGEMENT

The authors would like to thank Dr. Matthias Passlack of Freescale Semiconductor Inc. for the experimental results and the helpful discussions.